

GA109 / GA209 SUPERCAPACITOR

Datasheet Revision 4.3, July 2018

This Datasheet should be read in conjunction with the CAP-XX Supercapacitors Product Guide which contains information common to our product lines.

Electrical Specifications

The GA109 is a single cell supercapacitor. The GA209 is a dual cell supercapacitor with two GA109 cells in series, so GA209 capacitance = Capacitance of GA109/2 and GA209 ESR = 2 x GA109 ESR.

Table 1: Absolute Maximum Ratings

Parameter	Name		Conditions	Min	Max	Units
Terminal	V _{peak}	GA109		0	2.75	V
Voltage		GA209			5.5	
Temperature	T _{max}			-40	+70	°C

Table 2: Electrical Characteristics

Parameter	Name		Conditions	Min	Typical	Max	Units
Terminal	V	GA109		0		2.5	V
Voltage	Vn	GA209		0		5.0	v
Canacitanas	С -	GA109	DC, 23°C	106	170	204	mF
Capacitance		GA209		72	90	108	
ESR	ESR	GA109	DC, 23°C		50	60	mΩ
ESK		GA209			95	114	
Leakage Current	I_L		2.3V, 23°C, 120hrs		1	2	μΑ
RMS Current	I _{RMS}		23°C			4.5	A
Peak Current ¹	I_P		23°C			30	A

¹Non-repetitive current, single pulse to discharge fully charged supercapacitor.

Table 3: Thickness

GA109F		No adhesive tape on underside of the supercapacitor	GA109G		Adhesive tape on underside, release tape removed
GA209F	2.1mm		GA209G	2.2mm	



Definition of Terms

In its simplest form, the Equivalent Series Resistance (ESR) of a capacitor is the real part of the complex impedance. In the time domain, it can be found by applying a step discharge current to a charged cell as in Fig. 1. In this figure, the supercapacitor is pre-charged and then discharged with a current pulse, I=1A for duration 0.01 secs.

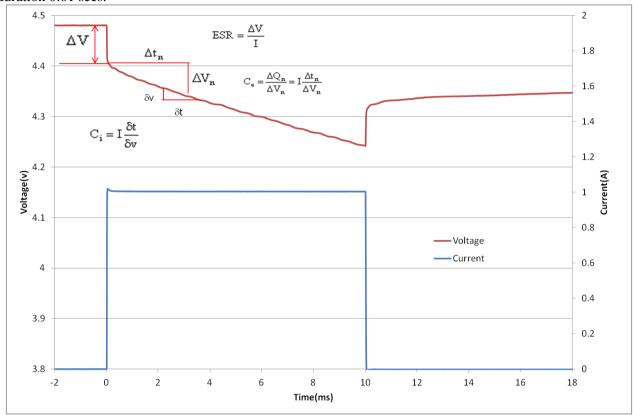


Figure 1: Effective capacitance, instantaneous capacitance and ESR for a GA209

The ESR is found by dividing the instantaneous voltage step (ΔV) by I. In this example = (4.48V-4.409V)/1.01A = 71m Ω .

The instantaneous capacitance (C_i) can be found by taking the inverse of the derivative of the voltage, and multiplying it by I.

The effective capacitance for a pulse of duration Δt_n , $Ce(\Delta t_n)$ is found by dividing the total charge removed from the capacitor (ΔQ_n) by the voltage lost by the capacitor (ΔV_n). For constant current $Ce(\Delta t_n) = I \ x$ $\Delta t_n/\Delta V_n$. Ce increases as the pulse width increases and tends to the DC capacitance value as the pulse width becomes very long (~10 secs). After 2msecs, Fig 1 shows the voltage drop $V_{2ms} = (4.409 V - 4.359 V) = 50 mV$. Therefore $Ce(2ms) = 1A \ x \ 2ms/50 mV = 40 mF$. After 10ms, the voltage drop = 4.409 V - 4.24V = 169 mV. Therefore $Ce(10ms) = 1A \ x \ 10 ms/169 mV = 59 mF$. The DC capacitance of a GA209 = 80 mF. Note that ΔV , or IR drop, is not included because very little charge is removed from the capacitor during this time. Ce shows the time response of the capacitor and it is useful for predicting circuit behavior in pulsed applications.



Measurement of DC Capacitance

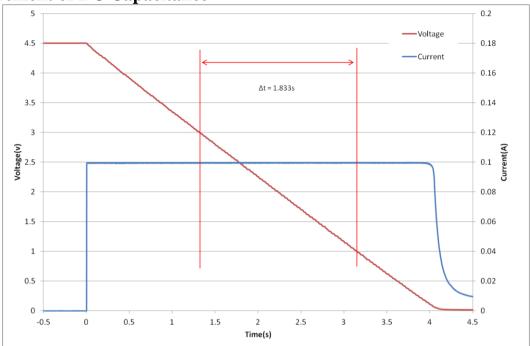


Fig 2: Measurement of DC Capacitance for a GA209

Fig 2 shows the measurement of DC capacitance by drawing a constant 100mA current from a fully charged supercapacitor and measuring the time taken to discharge from 1.5V to 0.5V for a single cell, or from 3V to 1V for a dual cell supercapacitor. In this case, $C = 0.1A \times 1.83 \text{ s}/2V = 91.5 \text{mF}$, which is well within the 90mF +/- 20% tolerance for a GA209 cell.

Measurement of ESR

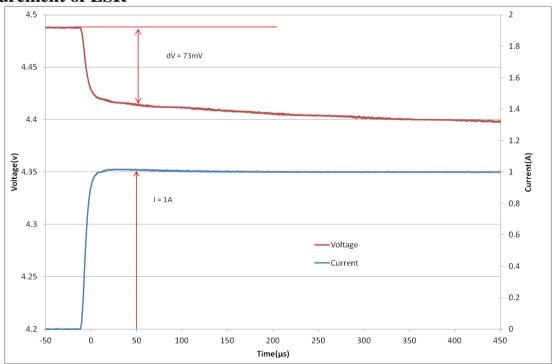


Fig 3: Measurement of ESR for a GA209

Fig 3 shows DC measurement of ESR by applying a step load current to the supercapacitor and measuring the resulting voltage drop. CAP-XX waits for a delay of $50\mu s$ after the step current is applied to ensure the voltage and current have settled. In this case the ESR is measured as $73\text{mV}/1A = 73\text{m}\Omega$.



Effective Capacitance

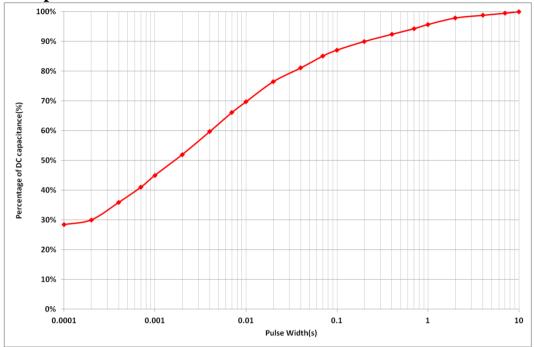


Figure 4: Effective Capacitance

Fig 4 shows the effective capacitance for the GA109, GA209 @ 23°C. This shows that for a 1msec PW, you will measure 45% of DC capacitance or 72mF for a GA109 or 36mF for a GA209. At 10msecs you will measure 70% of the DC capacitance, and at 100msecs you will measure 87% of DC capacitance. Ceffective is a time domain representation of the supercapacitor's frequency response. If, for example, you were calculating the voltage drop if the supercapacitor was supporting 1A for 10msecs, then you would use the Ceff(10msecs) = 70% of DC capacitance = 56mF for a GA209, so Vdrop = 1A x ESR + 1A x duration/C = 1A x 130m Ω + 1A x 10ms / 56mF = 308.6mV. The next section on pulse response shows how the effective capacitance is sufficient for even short pulse widths.

Pulse Response

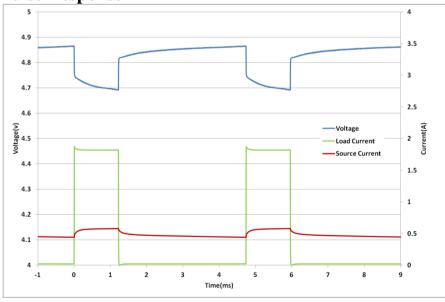


Fig 5 shows that the GA209 supercapacitor does an excellent job supporting a GPRS class 10 pulse train, drawing 1.8A for 1.1ms at 25% duty cycle. The source is current limited to 0.6A and the supercapacitor provides the 1.2A difference to achieve the peak current. At first glance the freq response of Fig 8 indicates the supercapacacitor would not support a 1ms pulse, but the Ceff of 36mF coupled with the low ESR supports this pulse train with only ~176mV droop in the supply rail.

Fig 5: GA209 Pulse Response with GPRS Class 10 Pulse Train



DC Capacitance variation with temperature

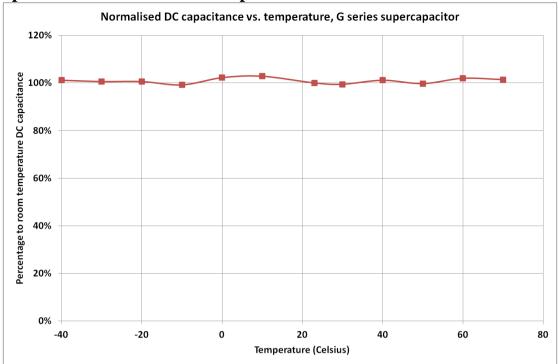


Figure 6: Capacitance change with temperature

Fig 6 shows that DC capacitance is approximately constant with temperature.

ESR variation with temperature

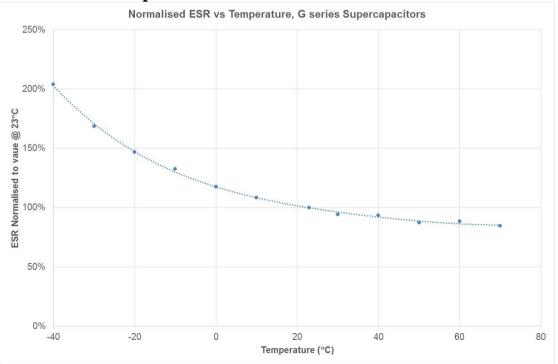


Figure 7: ESR change with temperature

Fig 7 shows that ESR at -40°C is \sim 2 x ESR at room temp, and that ESR at 70°C is \sim 0.85 x ESR at room temperature.



Frequency Response

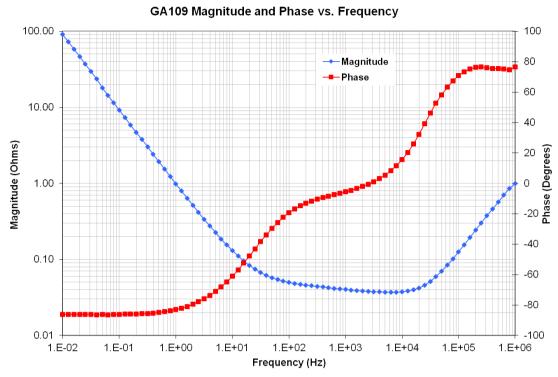


Fig 8: Frequency Response of Impedance (biased at 2.3V with a 50mV test signal)

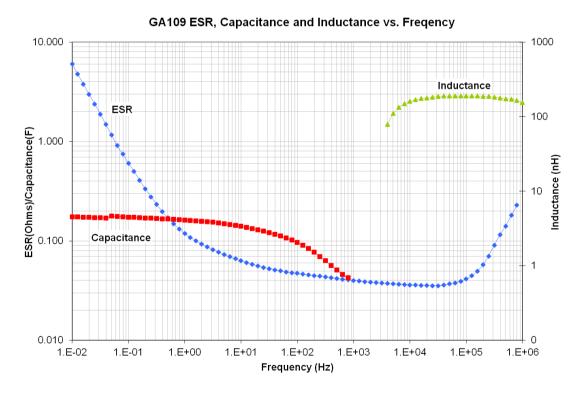


Fig 9: Frequency Response of ESR, Capacitance & Inductance

Fig 8 shows the supercapacitor behaves as an ideal capacitor until approx 30 Hz when the magnitude no longer rolls off proportionally to 1/freq and the phase crosses -45°. Performance of supercapacitors with frequency is complex and the best predictor of performance is Fig 4 showing effective capacitance as a function of pulsewidth.



Leakage Current

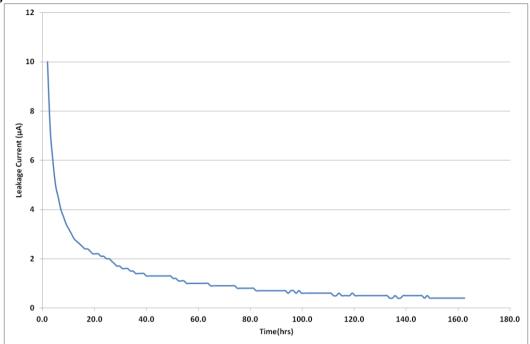


Fig 10: Leakage Current

Fig 10 shows the leakage current for GA109 at room temperature. The leakage current decays over time and the equilibrium value leakage current will be reached after ~120hrs at room temperature. The typical equilibrium leakage current is $0.5\mu A$ at room temperature. At $70^{\circ}C$ leakage current will be ~ $5\mu A$.



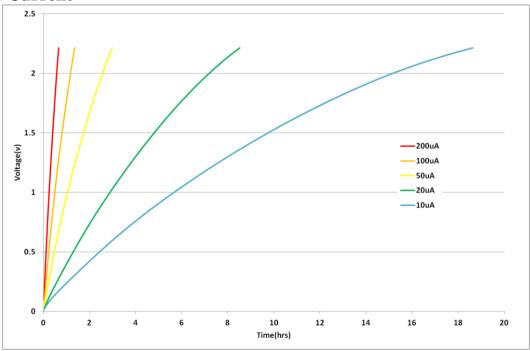


Fig 11: Charging an GA109 with low current

The corollary to the slow decay in leakage currents shown in Fig 10 is that charging a supercapacitor at very low currents takes longer than theory predicts. At higher charge currents, the charge rate is as theory predicts. For example, it should take 160mF x 2.3V / 0.00001A = 10.2hrs to charge a 160mF supercapacitor to 2.3V at $10\mu A$, but Fig 11 shows it took 19hrs. At $50\mu A$ charging occurs at a rate close to the theoretical rate.



RMS Current

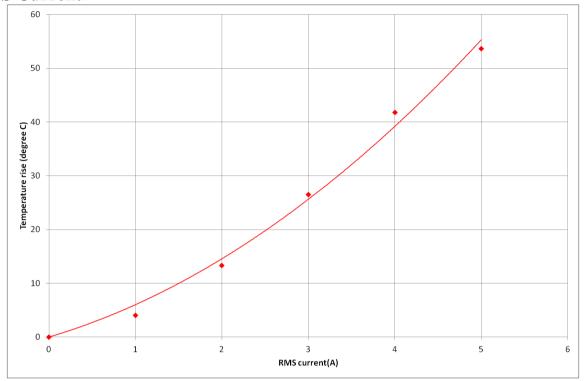


Fig 12: Temperature rise in GA209 with RMS current

Continuous current flow into/out of the supercap will cause self heating, which limits the maximum continuous current the supercapacitor can handle. This is measured by a current square wave with 50% duty cycle, charging the supercapacitor to rated voltage at a constant current, and then discharging the supercapacitor to half rated voltage at the same constant current value. For a square wave with 50% duty cycle, the RMS current is the same as the current amplitude. Fig 12 shows the increase in temperature as a function of RMS current. From this, the maximum RMS current in an application can be calculated, for example, if the ambient temperature is 40°C, and the maximum desired temperature for the supercapacitor is 70°C, then the maximum RMS current should be limited to 3.3 A, which causes a 30°C temperature increase.

CAP-XX Supercapacitors Product Guide

Refer to the package drawings in the CAP-XX Supercapacitors Product Guide for detailed information of the product's dimensions, PCB landing placements, active areas and electrical connections.

Refer to the CAP-XX Supercapacitors Product Guide for information on endurance and shelf life, transportation and storage, assembly and soldering, safety and RoHS/EREACH certification.



GA130 / GA230 SUPERCAPACITOR

Datasheet Rev 4.3, July 2018

This Datasheet should be read in conjunction with the CAP-XX Supercapacitors Product Guide which contains information common to our product lines.

Electrical Specifications

The GA130 is a single cell supercapacitor. The GA230 is a dual cell supercapacitor with two GA130 cells in series, so GA230 capacitance = Capacitance of GA130/2 and GA230 ESR = 2 x GA130 ESR.

Table 1: Absolute Maximum Ratings

Parameter	Name		Conditions	Min	Typical	Max	Units
Terminal	Vpeak	GA130		0		2.75	V
Voltage		GA230				5.5	
Temperature	Tmax			-40		+70	°C

Table 2: Electrical Characteristics

Parameter	Name		Conditions	Min	Typical	Max	Units
Terminal	Vn	GA130		0		2.5	V
Voltage	VII	GA230		0		5.0	
Capacitance	С	GA130	DC, 23°C	640	800	960	mF
Capacitance		GA230		320	400	480	
ESR	ESR	GA130	DC, 23°C		40	48	mΩ
ESK		GA230			75	90	
Leakage Current	I_L		2.3V, 23°C 120hrs		1	2	μΑ
RMS Current	I_{RMS}		23°C			4.5	A
Peak Current ¹	I_P		23°C			30	A

Non-repetitive current, single pulse to discharge fully charged supercapacitor.

Table 3: Thickness

GA130F	1.7mm	No adhesive tape on underside of the supercapacitor	GA130G		Adhesive tape on underside, release tape removed
GA230F	3.5mm		GA230G	3.6mm	



Definition of Terms

In its simplest form, the Equivalent Series Resistance (ESR) of a capacitor is the real part of the complex impedance. In the time domain, it can be found by applying a step discharge current to a charged cell as in Fig. 1. In this figure, the supercapacitor is pre-charged and then discharged with a current pulse, I=1A for duration 0.01 sec.

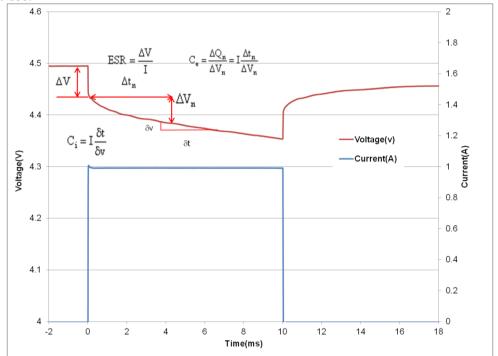


Figure 1: Effective capacitance, instantaneous capacitance and ESR for a GA230

The ESR is found by dividing the instantaneous voltage step (ΔV) by I. In this example = $(4.49V-4.43V)/1A = 60m\Omega$.

The instantaneous capacitance (C_i) can be found by taking the inverse of the derivative of the voltage, and multiplying it by I.

The effective capacitance for a pulse of duration Δt_n , $Ce(\Delta t_n)$ is found by dividing the total charge removed from the capacitor (ΔQ_n) by the voltage lost by the capacitor (ΔV_n). For constant current $Ce(\Delta t_n) = I \ x$ $\Delta t_n/\Delta V_n$. Ce increases as the pulse width increases and tends to the DC capacitance value as the pulse width becomes very long (~10 secs). After 2msecs, Fig 1 shows the voltage drop $V_{2ms} = (4.43V - 4.399V) = 31mV$. Therefore $Ce(2ms) = 1A \ x \ 2ms/31mV = 64.5mF$. After 10ms, the voltage drop = 4.43 V - 4.353V = 77mV. Therefore $Ce(10ms) = 1 \ A \ x \ 10ms/77mV = 129.9mF$. The DC capacitance of a GA230 = 0.4 F. Note that ΔV , or IR drop, is not included because very little charge is removed from the capacitor during this time. Ce shows the time response of the capacitor and it is useful for predicting circuit behavior in pulsed applications.

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Measurement of DC Capacitance

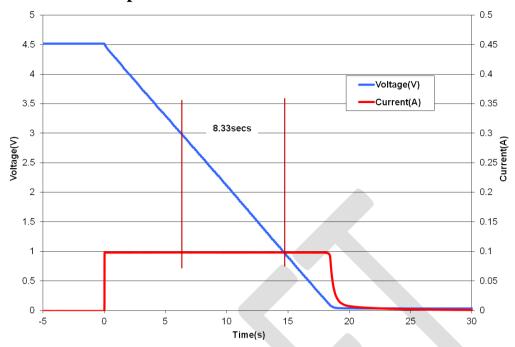


Fig 2: Measurement of DC Capacitance for a GA230

Fig 2 shows the measurement of DC capacitance by drawing a constant 100mA current from a fully charged supercapacitor and measuring the time taken to discharge from 1.5V to 0.5V for a single cell, or from 3V to 1V for a dual cell supercapacitor. In this case, $C = 0.1A \times 8.33s / 2V = 416.5mF$, which is well within the 400mF + 20% tolerance for a GA230 cell.

Measurement of ESR

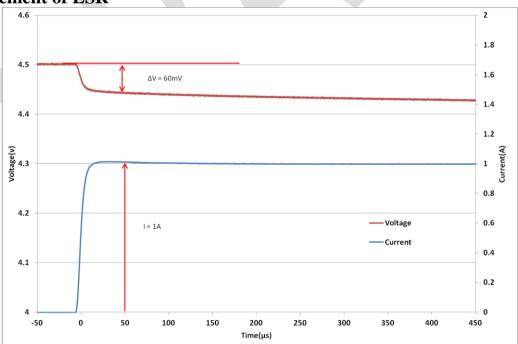


Fig 3: Measurement of ESR for a GA230

Fig 3 shows DC measurement of ESR by applying a step load current to the supercapacitor and measuring the resulting voltage drop. CAP-XX waits for a delay of $50\mu s$ after the step current is applied to ensure the voltage and current have settled. In this case the ESR is measured as $60mV/1A = 60m\Omega$.



Effective Capacitance

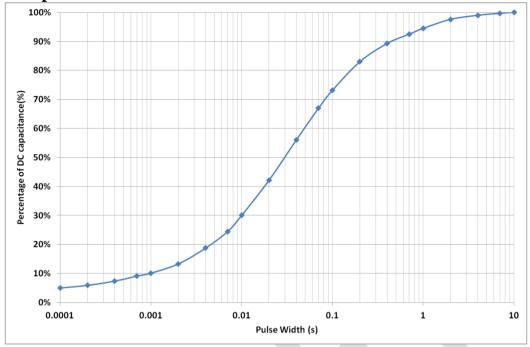
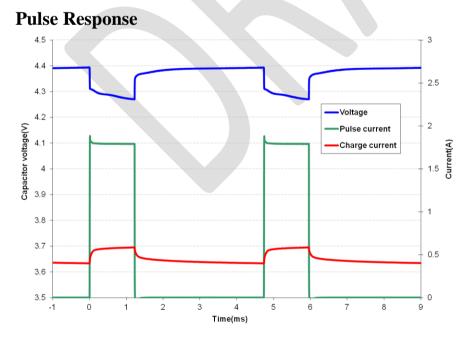


Figure 4: Effective Capacitance

Fig 4 shows the effective capacitance for the GA130, GA230 @ 23°C. This shows that for a 1msec PW, you will measure 10.1% of DC capacitance or 80.8mF for a GA130 or 40.4mF for a GA230. At 10msecs you will measure 30.8% of the DC capacitance, and at 100msecs you will measure 74% of DC capacitance. Ceffective is a time domain representation of the supercapacitor's frequency response. If, for example, you were calculating the voltage drop if the supercapacitor was supporting 1A for 10msecs, then you would use the Ceff(10msecs) = 30.8% of DC capacitance = 123.2mF for an GA230, so Vdrop = 1A x ESR + 1A x duration/C = 1A x $60m\Omega + 1A \times 10ms / 123.2mF = 141.2mV$. The next section on pulse response shows how the effective capacitance is sufficient for even short pulse widths.



supercapacitor provides the 1.2A difference to achieve the peak current. At first glance the freq response of Fig 8 indicates the supercapacacitor would not support a 1ms pulse, but the Ceff of 40mF coupled with the low ESR supports this pulse train with only ~120mV droop in the supply rail.

Fig 5 shows that the GA230

supercapacitor does an

excellent job supporting a

GPRS class 10 pulse train,

drawing 1.8A for 1.1ms at

25% duty cycle. The source is

current limited to 0.6A and the

Fig 5: GA230 Pulse Response with GPRS Class 10 Pulse Train



DC Capacitance variation with temperature

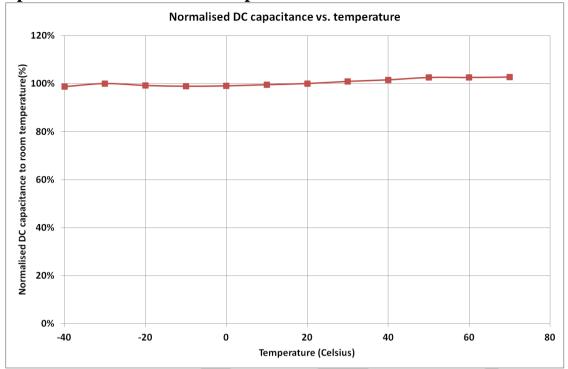


Figure 6: Capacitance change with temperature

Fig 6 shows that DC capacitance is approximately constant with temperature.

ESR variation with temperature

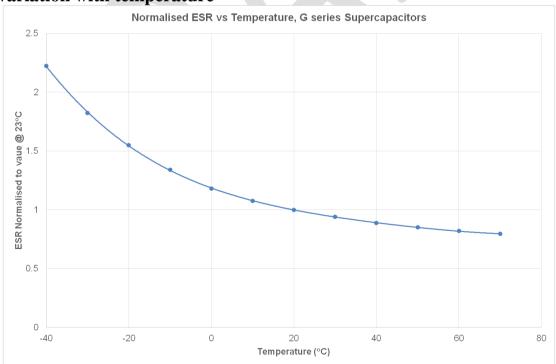


Figure 7: ESR change with temperature

Fig 7 shows that ESR at -40°C is \sim 2.2 x ESR at room temp, and that ESR at 70°C is \sim 0.8 x ESR at room temperature.



Frequency Response

GA130 Magnitude and Phase vs. Frequency

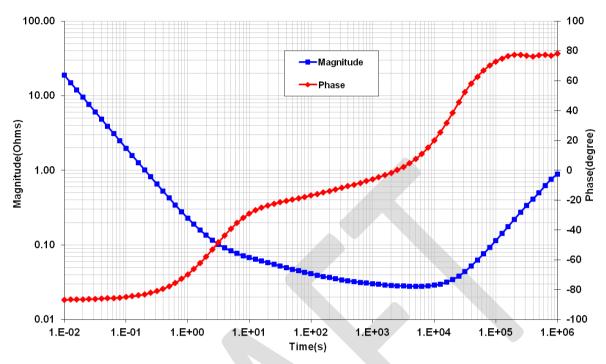


Fig 8: Frequency Response of Impedance (biased at 2.3V with a 50mV test signal)

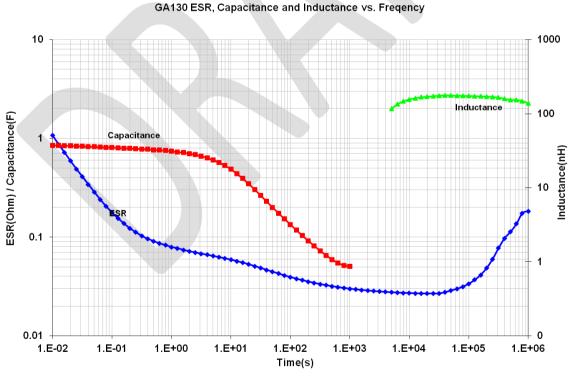


Fig 9: Frequency Response of ESR, Capacitance & Inductance

Fig 8 shows the supercapacitor behaves as an ideal capacitor until approx 20 Hz when the magnitude no longer rolls off proportionally to 1/freq and the phase crosses - 45° . Performance of supercapacitors with frequency is complex and the best predictor of performance is Fig 4 showing effective capacitance as a function of pulsewidth.



Leakage Current

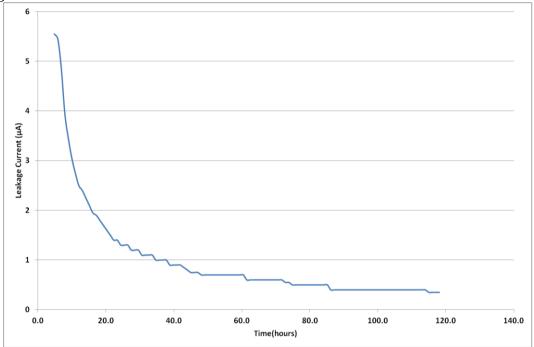


Fig 10: Leakage Current

Fig 10 shows the leakage current for GA130 at room temperature. The leakage current decays over time, and the equilibrium value leakage current will be reached after ~120hrs at room temperature. The typical equilibrium leakage current is $0.5\mu A$ at room temperature. At $70^{\circ}C$ leakage current will be ~ $5\mu A$.



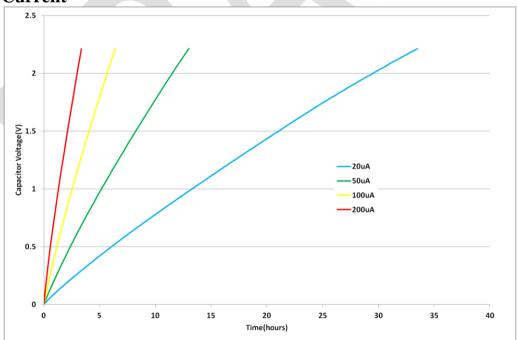


Fig 11: Charging an GA130 with low current

The corollary to the slow decay in leakage currents shown in Fig 10 is that charging a supercapacitor at very low currents takes longer than theory predicts. At higher charge currents, the charge rate is as theory predicts. For example, it should take $0.8F \times 2.4V / 0.00002A = 26.7hrs$ to charge a 0.8F supercapacitor to 2.4V at $20\mu A$, but Fig 11 shows it took 33hrs. At $100\mu A$ charging occurs at a rate close to the theoretical rate.



RMS Current

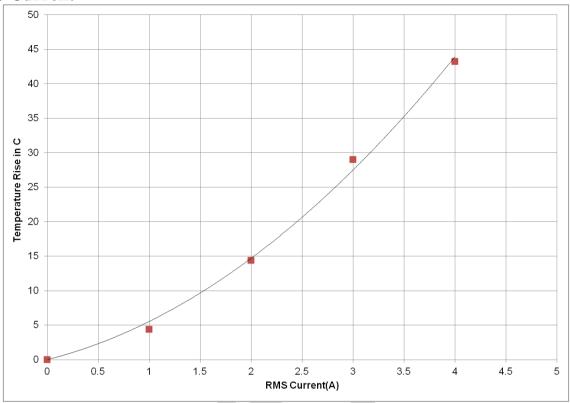


Fig 12: Temperature rise in GA230 with RMS current

Continuous current flow into/out of the supercap will cause self heating, which limits the maximum continuous current the supercapacitor can handle. This is measured by a current square wave with 50% duty cycle, charging the supercapacitor to rated voltage at a constant current, then discharging the supercapacitor to half rated voltage at the same constant current value. For a square wave with 50% duty cycle, the RMS current is the same as the current amplitude. Fig 12 shows the increase in temperature as a function of RMS current. From this, the maximum RMS current in an application can be calculated, for example, if the ambient temperature is 40°C, and the maximum desired temperature for the supercapacitor is 70°C, then the maximum RMS current should be limited to 3A, which causes a 30°C temperature increase.

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